Efficient Model-Free Learning to Overcome Hardware Nonidealities in Analog-to-Information Converters

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Multiband RF Spectrum Sensing

Spectrum sensing: efficiently determine the frequency content of an unknown real signal x(t).



Assume there are n bands of width B in the partition.

Classical Spectrum Sensing Receiver Designs



With ideal components, a linear system: y(f) = Ax(f)

To first order, the receivers have equal energy consumption.

Compressed Sensing Receivers for Spectrum Sensing

Compressed sensing¹: invert underdetermined y = Ax via

- Structured input x,
- Generic sensor A,
- Convex optimization for reconstruction.



 $\begin{array}{ll} \underset{\boldsymbol{x} \in \mathbb{R}^n}{\text{minimize}} & \|\boldsymbol{x}\|_1 \\ \text{subject to} & \|\boldsymbol{y} - \boldsymbol{A} \boldsymbol{x}\|_2^2 \leq \delta \end{array}$



A recipe from numerical optimization: solve

$$\min_{\boldsymbol{x} \in \mathbb{C}^n} \lambda \|\boldsymbol{x}\|_1 + \frac{1}{2} \|\boldsymbol{y} - \boldsymbol{A}\boldsymbol{x}\|_2^2$$

via the proximal gradient method:

$$oldsymbol{z}^{(k)} \leftarrow oldsymbol{x}^{(k)} - rac{1}{L}oldsymbol{A}^*(oldsymbol{A}oldsymbol{x}^{(k)} - oldsymbol{y})
onumber \ oldsymbol{x}^{(k+1)} \leftarrow e^{\hat{\mathfrak{i}} \arg(oldsymbol{z}^{(k)})} \max\{\sum_i |z_i^{(k)}| oldsymbol{e}_i - rac{\lambda}{L} oldsymbol{1}, oldsymbol{0}\}$$

for
$$k = 0, 1, ..., L = ||\mathbf{A}||^2$$
, $\mathbf{x}^{(0)} = \mathbf{0}$ (say).

Characteristics of numerical optimization approaches to the recovery problem:

- Targeted towards **general** classes of problems.
- Accompanied by **worst-case** performance guarantees.
- Hand-designed using known models for the application.

$$\inf \underbrace{f}_{(\text{smooth})} + \underbrace{g}_{(\text{convex})} = h$$

$$h(\boldsymbol{x}^{(k)}) - h(\boldsymbol{x}_{\star}) = O(1/k)$$

$$x ext{ sparse } \implies g = \|\cdot\|_1$$

On the other hand, for spectrum sensing, we would like an algorithm that satisfies:

subclass of sparse recovery problems.



- Able to incorporate hard constraints on computational resources.
- Adaptive to deviations from the nominal design in a model-free manner.



A neural network consists of:

- 1. An integer $d \ge 1$ (the **depth**)
- 2. Euclidean spaces \mathbb{R}^{n_i} , $i = 0, \dots, d$
- 3. Affine maps $\mathcal{L}_i : \mathbb{R}^{n_{i-1}} \to \mathbb{R}^{n_i}$, $i \in [d]$
- 4. "Nonlinearities" $\sigma_i : \mathbb{R}^{n_i} \to \mathbb{R}^{n_i}, i \in [d]$

Iteratively: define

$$f^{(k)} = \begin{cases} \sigma_k(\mathcal{L}_k(f^{(k-1)})) & k \ge 1\\ I_{n_0} & k = 0 \end{cases}$$

Then $\boldsymbol{y} = f^{(d)}(\boldsymbol{x})$.

For spectrum sensing: which architecture do we choose?



Recall the iteration

$$oldsymbol{z}^{(k)} \leftarrow oldsymbol{x}^{(k)} - rac{1}{L}oldsymbol{A}^*(oldsymbol{A}oldsymbol{x}^{(k)} - oldsymbol{y})$$

 $oldsymbol{x}^{(k+1)} \leftarrow e^{\hat{\mathfrak{i}} \arg(oldsymbol{z}^{(k)})} \max\{\sum_i |z_i^{(k)}| oldsymbol{e}_i - rac{\lambda}{L} oldsymbol{1}, oldsymbol{0}\}$

Define

$$p_{\lambda}(\boldsymbol{x}) = e^{\hat{i} \arg(\boldsymbol{x})} \max\{\sum_{i} |x_i| \boldsymbol{e}_i - \frac{\lambda}{L} \mathbf{1}, \mathbf{0}\}.$$

Then after rearranging:

$$\boldsymbol{x}^{(k+1)} \leftarrow p_{\lambda} \left(\frac{1}{L} \boldsymbol{A}^* \boldsymbol{y} + (\boldsymbol{I} - \frac{1}{L} \boldsymbol{A}^* \boldsymbol{A}) \boldsymbol{x}^{(k)} \right),$$

which has the form of a neural network.



Truncate the network, and **learn** its parameters using data. The result satisfies:

- 1. Optimized for a **specific** computational budget and sparse inverse problem.
- Adaptive to nonidealities via the training procedure, without explicit modeling.
- 3. Leverages available **prior information** via the network topology and initializations.
- This approach is called LISTA [Gregor and LeCun 2010].



Other works on learned optimization algorithms:

- (Near-)SoA in super-resolution [Wang et al. 2015], voice identification [Sprechmann, A. Bronstein, et al. 2013]
- Additional structural priors [Sprechmann, A. M. Bronstein, and Sapiro 2015]
- Other base recovery algorithms: AMP [Borgerding, Schniter, and Rangan 2016], SBL [He, Xin, and Wipf 2017]
- Co-optimization of nonlinearities and weights [Kamilov and Mansour 2015; Mahapatra, Mukherjee, and Seelamantula 2017]

Learned Optimization Training Procedure

Reparametrize the network. Then perform empirical risk minimization

$$\underset{\boldsymbol{W},\boldsymbol{S}}{\text{minimize}} \quad \frac{1}{2} \sum_{i=1}^{T} \left\| \boldsymbol{x}^{(d)}(\boldsymbol{y}_{i}) - \overline{\boldsymbol{x}_{i}} \right\|_{2}^{2}$$

using

1. a dataset $\{oldsymbol{y}_i, \overline{oldsymbol{x}}_i\}_{i=1}^T$

2. initializations based on the design matrix \widetilde{A} to obtain (W_{\star}, S_{\star}) . We desire that:

- 1. $T \mbox{ is not too large}$
- 2. Recovery guarantees associated with \widetilde{A} are associated with (W_\star, S_\star)



LISTA for Multiband Spectrum Sensing - Training Set



Proposed dataset: measurements y and corresponding spectra x for each possible 1-sparse x(t), at moderate power.

• Compare to

$$\sum_{k=1}^{\lceil m/\log n\rceil} \binom{n}{k}$$

signals for exhaustive training.

Test on arbitrary k-sparse signals with equal support powers.

Experiments: Simulation Setup



We simulate the DRF2IC receiver [Haque et al. 2018], and $A \in \mathbb{C}^{18 \times 63}$.

We consider two realistic receiver nonidealities:

- 1. IQ downconversion frequency-independent mismatch
- 2. PRBS fractional phase offsets

Task: support recovery; evaluation metric: P_{D} and P_{FA}

Experiments: Detection vs. SNR



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Experiments: Detection vs. Sparsity



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Experiments: Test Set NMSE vs. Training Parameters



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- Learned optimization-based signal recovery methodology was argued to be **ideal** for interferer detection in CS-based spectrum sensing receivers.
- 2. An efficient training protocol was developed and evaluated in the equal-power and linearly-impaired regime.
- 3. Extensions to more challenging signal models and receiver impairments will require additional innovations.





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Thanks! Questions?

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A recipe from numerical optimization: solve

$$\underset{\boldsymbol{X} \in \mathbb{C}^{n \times p}}{\text{minimize}} \quad \lambda \sum_{i=1}^{n} \|\boldsymbol{x}_{i}\|_{2} + \frac{1}{2} \|\boldsymbol{Y} - \boldsymbol{A}\boldsymbol{X}\|_{F}^{2}$$

via the proximal gradient method:

$$oldsymbol{Z}^{(k)} \leftarrow oldsymbol{X}^{(k)} - rac{1}{L}oldsymbol{A}^*(oldsymbol{A}oldsymbol{X}^{(k)} - oldsymbol{Y}) \ oldsymbol{X}^{(k+1)} \leftarrow e^{\hat{\mathfrak{i}} rg(oldsymbol{Z}^{(k)})} \max\{|oldsymbol{Z}^{(k)}| - rac{\lambda}{L}\sum_{i=1}^n \|oldsymbol{x}_i\|oldsymbol{e}_i \otimes oldsymbol{1}_p, oldsymbol{0}\}$$

for $k=1,2,\ldots$, $L=\|\boldsymbol{A}\|^2$, $\boldsymbol{X}^{(0)}=\boldsymbol{0}$ (say).

All experiments focus on the MMV problem

- 1. $f_{min} = 2.57 \text{ GHz}, f_{max} = 3.83 \text{ GHz}$
- 2. Sampling frequency set for an OSR of 8
- 3. DRF2IC run with 2 physical branches and expansion factor of 9

4.
$$m = 18$$
, $n = 63$, $p = 80$

- 5. IQ imbalance of 3 dB, 30°
- 6. PN sequence phase misalignment of 6/8

LISTA parameters: d = 12, T = 63, $\lambda = 0.1$. Data is labeled via a pseudoinverse on the known support.

Support recovery details:

- 1. OMP: terminate after k + 2 iterations
- 2. ℓ^1 and LISTA: compute the average bin power in the predicted signal, and declare the top k + 2 bins to be active

- 1. Number of samples per point: 200
- 2. Sparsity levels: 3, 6
- 3. Samples are uniformly random k-sparse signals, plus independent noise
- 4. Noise power calculated over the entire band
- 5. Orthogonal matching pursuit run with only the previously-mentioned stopping criterion
- 6. ℓ^1 implementation is an accelerated proximal gradient algorithm
- 7. Calibrated traces run the corresponding algorithms on unimpaired data

- 1. Number of samples per point: 500
- 2. SNR is fixed at 20 dB
- 3. Signal generation and algorithms as in the previous experiment

- 1. Number of samples per point: 500
- 2. NMSE calculated similarly to the LISTA training objective (just with normalization)
- 3. SNR is fixed at 20 dB, sparsity is fixed at k=6
- 4. Signal generation as in the previous experiment